

## Six-Channel Class AB Audio Power Amplifier IC 30 W $\times 6$ Channels

## Preliminary

## Overview

The STK403-400 series products are audio power amplifier hybrid ICs that consist of optimally-designed discrete component power amplifier circuits that have been miniaturized using SANYO's unique insulated metal substrate technology (IMST). The adoption of a newlydeveloped low thermal resistance substrate allows this product to integrate six power amplifier channels in a single compact package. The adoption of a standby circuit in this device allows it to reduce impulse noise significantly as compared to earlier Sanyo products, in particular, the STK402-*00 series products.

## Features

- Series of pin compatible power amplifiers ranging from $30 \mathrm{~W} / \mathrm{ch}$ to $45 \mathrm{~W} / \mathrm{ch}(10 \% / 1 \mathrm{kHz})$ devices. The same printed circuit board can be used depending on the output power grade.
- Miniature packages
$-78.0 \mathrm{~mm} \times 32.0 \mathrm{~mm} \times 9.0 \mathrm{~mm}$ *
*: Not including the pins.
- Output load impedance: $\mathrm{R}_{\mathrm{L}}=6 \Omega$
- Allowable load shorted time: 0.3 seconds
- Supports the use of standby and muting circuits.


## Package Dimensions

unit: mm

4202-SIP28

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## Series Organization

These products are organized as a series based on their output capacity.

| Item | Type No. |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
|  | STK403-430 | STK403-440 | STK403-450 |  |  |
| Output $1(10 \% / 1 \mathrm{kHz})$ | $30 \mathrm{~W} \times 6 \mathrm{ch}$ | $40 \mathrm{~W} \times 6 \mathrm{ch}$ | $45 \mathrm{~W} \times 6 \mathrm{ch}$ |  |  |
| Output $2(0.6 \% / 20 \mathrm{~Hz}$ to 20 kHz$)$ | $20 \mathrm{~W} \times 6 \mathrm{ch}$ | $25 \mathrm{~W} \times 6 \mathrm{ch}$ | $30 \mathrm{~W} \times 6 \mathrm{ch}$ |  |  |
| Maximum supply voltage (No signal) | $\pm 36 \mathrm{~V}$ | $\pm 38 \mathrm{~V}$ | $\pm 40 \mathrm{~V}$ |  |  |
| Maximum supply voltage $(6 \Omega)$ | $\pm 34 \mathrm{~V}$ | $\pm 36 \mathrm{~V}$ | $\pm 38 \mathrm{~V}$ |  |  |
| Recommended supply voltage $(6 \Omega)$ | $\pm 23 \mathrm{~V}$ | $\pm 26 \mathrm{~V}$ | $\pm 28 \mathrm{~V}$ |  |  |
| Package | $78.0 \mathrm{~mm} \times 32.0 \mathrm{~mm} \times 9.0 \mathrm{~mm}$ |  |  |  |  |

## Specifications

## Maximum Ratings at $\mathbf{T a}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Maximum supply voltage (No signal) | $\mathrm{V}_{\text {CC }} \max (0)$ |  | $\pm 40$ | V |
| Maximum supply voltage | $\mathrm{V}_{C C} \max (1)$ | $\mathrm{R}_{\mathrm{L}}=6 \Omega$ | $\pm 38$ | V |
| Minimum operating supply voltage | $\mathrm{V}_{\mathrm{CC}}$ min |  | $\pm 10$ | V |
| Maximum operation flow-in current (pin 23) | IST OFF max |  | 1.2 | mA |
| Thermal resistance | $\theta \mathrm{j}$-c | Per power transistor | 3.6 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| Junction temperature | Tj max | Both the Tj max and the Tc max conditions must be met. | 150 | ${ }^{\circ} \mathrm{C}$ |
| Operating IC substrate temperature | Tc max |  | 125 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  | -30 to +125 | ${ }^{\circ} \mathrm{C}$ |
| Allowable load shorted time *4 | ts | $\mathrm{V}_{\mathrm{CC}}= \pm 28.0 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=6 \Omega, \mathrm{f}=50 \mathrm{~Hz}, \mathrm{P}_{\mathrm{O}}=30 \mathrm{~W}, 1 \mathrm{ch}$ drive | 0.3 | s |

Operating Characteristics at $\mathrm{Tc}=25^{\circ} \mathrm{C}, \mathrm{R}_{\mathrm{L}}=6 \Omega$ (noninductive load), $\mathrm{Rg}=\mathbf{6 0 0} \Omega, \mathrm{VG}=30 \mathrm{~dB}$

| Parameter | Symbol | Conditions*1 |  |  |  |  | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\mathrm{V}_{\mathrm{CC}}$ (V) | $\mathrm{f}(\mathrm{Hz})$ | Po (W) | THD (\%) |  | min | typ | max |  |
| Output power *1 | $\mathrm{P}_{\mathrm{O}}$ (1) | $\pm 28.0$ | 20 to 20 k |  | 0.6 |  | 27 | 30 |  | W |
|  | Po (2) | $\pm 28.0$ | 1 k |  | 10 |  |  | 45 |  |  |
| Total harmonic distortion *1 | THD (1) | $\pm 28.0$ | 20 to 20 k | 5.0 |  | $\mathrm{VG}=30 \mathrm{~dB}$ |  |  | 0.6 | \% |
|  | THD (2) | $\pm 28.0$ | 1 k | 5.0 |  | $\mathrm{VG}=30 \mathrm{~dB}$ |  | 0.03 |  |  |
| Frequency characteristics | $\mathrm{f}_{\mathrm{L},} \mathrm{f}_{\mathrm{H}}$ | $\pm 28.0$ |  | 1.0 |  | +0-3 dB | 20 to 50 k |  |  | Hz |
| Input impedance | ri | $\pm 28.0$ | 1 k | 1.0 |  |  |  | 55 |  | k $\Omega$ |
| Output noise voltage *2 | $\mathrm{V}_{\mathrm{NO}}$ | $\pm 34.0$ |  |  |  | $\mathrm{Rg}=2.2 \mathrm{k} \Omega$ |  |  | 1.0 | mV rms |
| Quiescent current | Icco | $\pm 34.0$ |  |  |  | No loading | 60 | 110 | 180 | mA |
| Neutral voltage | $\mathrm{V}_{\mathrm{N}}$ | $\pm 34.0$ |  |  |  |  | -70 | 0 | +70 | mV |
| Current flowing into pin 23 in standby mode *6 | IST ON | $\pm 28.0$ |  |  |  | $\mathrm{V}_{23}=5 \mathrm{~V}$, current limiting resistance: $6.2 \mathrm{k} \Omega$ |  |  | 0 | mA |
| Current flowing into pin 23 in operating mode *6 | ISt OFF | $\pm 28.0$ |  |  |  |  | 0.4 |  | 1.2 | mA |

Notes: 1. 1ch drive
2. Unless otherwise noted, use a constant-voltage supply for the power supply used during inspection.
3. The output noise voltage values shown are peak values read with a VTVM. However, an AC stabilized ( 50 Hz ) power supply should be used to minimize the influence of AC primary side flicker noise on the reading.
4. Use the transformer power supply circuit shown in the figure below for allowable load shorted time measurement and output noise voltage measurement.
5. Design applications so that the minus pre- $\mathrm{V}_{\mathrm{CC}}$ line (pin 17) is at the lowest potential at all times.
6. A limiting resistor that assures that the maximum operating current flowing into the standby pin (pin 23) does not exceed the maximum rating must be included in application circuits. This IC operates when a voltage higher than $\mathrm{V}_{\mathrm{BE}}$ (about 0.6 V ) is applied to the standby pin.


Designated Transformer Power Supply (RP-25 equivalent)


## Sample Application Circuit


*: Use a value for the limiting resistor that assures that the maximum operating current flowing into the standby pin (pin 23 ) does not exceed the maximum rating.

## Thermal Design Example

The heat sink thermal resistance, $\theta \mathrm{c}-\mathrm{a}$, required to handle the total power dissipated within this hybrid IC is determined as follows.

Condition 1: The IC substrate temperature Tc must not exceed $125^{\circ} \mathrm{C}$.
$\mathrm{Pd} \times \theta \mathrm{c}-\mathrm{a}+\mathrm{Ta}<125^{\circ} \mathrm{C} \ldots$ (1)
Ta: Guaranteed ambient temperature for the end product.
Condition 2: The junction temperature of each transistor must not exceed $150^{\circ} \mathrm{C}$.
$\mathrm{Pd} \times \theta \mathrm{c}-\mathrm{a}+\mathrm{Pd} / \mathrm{N} \times \theta \mathrm{j}-\mathrm{c}+\mathrm{Ta}<150^{\circ} \mathrm{C} \ldots$ (2)
N : Number of power transistors
$\theta \mathrm{j}$-c: Thermal resistance per power transistor
We take the power dissipation in the power transistors to be Pd evenly distributed across those N power transistors.
If we solve for $\theta \mathrm{c}-\mathrm{a}$ in equations (1) and (2), we get the following inequalities.

$$
\begin{aligned}
& \theta \mathrm{c}-\mathrm{a}<(125-\mathrm{Ta}) / \mathrm{Pd} \ldots(1)^{\prime} \\
& \theta \mathrm{c}-\mathrm{a}<(150-\mathrm{Ta}) / \mathrm{Pd}-\mathrm{\theta}-\mathrm{c} / \mathrm{N} \ldots(2)^{\prime}
\end{aligned}
$$

Values that satisfy both these inequalities at the same time are the required heat sink thermal resistance values.
Determining the following specifications allows us to obtain the required heat sink thermal resistance from inequalities (1)' and (2)'.

- Supply voltage: $\mathrm{V}_{\mathrm{CC}}$
- Load resistance: $\mathrm{R}_{\mathrm{L}}$
- Guaranteed ambient temperature: Ta


## Example:

Assume that the IC supply voltage, $\mathrm{V}_{\mathrm{CC}}$, is $\pm 28 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}$ is $6 \Omega$, and that the signal is a continuous sine wave. In this case, from the $\mathrm{Pd}-\mathrm{P}_{\mathrm{O}}$ characteristics, the maximum power will be 164 W for a signal with a frequency of 1 kHz .
For actual music signals, it is usual to use a Pd of $1 / 8$ of $\mathrm{P}_{\mathrm{O}} \mathrm{max}$, which is the power estimated for continuous signals in this manner. (Note that depending on the particular safety standard used, a value somewhat different from the value of $1 / 8$ used here may be used.)

That is:
$\mathrm{Pd}=105 \mathrm{~W}$ (when $1 / 8 \mathrm{P}_{\mathrm{O}}$ max is 3.8 W )
The number, N , of power transistors in the hybrid IC's audio amplifier block is 12 . Since the thermal resistance, $\theta \mathrm{j}$-c, per transistor is $3.6^{\circ} \mathrm{C} / \mathrm{W}$, the required heat sink thermal resistance, $\theta \mathrm{c}-\mathrm{a}$, for a guaranteed ambient temperature of $50^{\circ} \mathrm{C}$ will be as follows.

From inequality (1)': $\theta \mathrm{c}-\mathrm{a}<(125-50) / 105$

$$
<0.71
$$

From inequality (2)': $\theta \mathrm{c}-\mathrm{a}<(150-50) / 105-3.6 / 12$

$$
<0.65
$$

Therefore, the thermal resistance that satisfies both these expressions at the same time is $0.65^{\circ} \mathrm{C} / \mathrm{W}$.
Note that this thermal design example assumes the use of a constant-voltage power supply, and is only provided as an example for reference purposes. Thermal designs must be tested in an actual end product.

## Stand-by \& Mute Sample Application Circuit



## Standby Mode Control



- Impulse noise that occurs at power on and power off can be reduced significantly by using a standby circuit.
- End product design is made easier by using a limiting resistor (*) to match the control voltage provided by the microcontroller or other control circuit.
- Standby control is available by controlling the current ( $I_{\text {ST }}$ ) flowing into the standby pin (pin 23).

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